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LM386 Low Voltage Audio Power Amplifier

National Semiconductor

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General Description

The LM386 is a power amplifier designed for use in low voltage consumer applications. The gain is internally set to 20 to keep external part count low, but the addition of an external resistor and capacitor between pins 1 and 8 will increase the gain to any value up to 200.

The inputs are ground referenced while the output is automatically biased to one half the supply voltage. The quiescent power drain is only 24 milliwatts when operating from a 6 volt supply, making the LM386 ideal for battery operation.

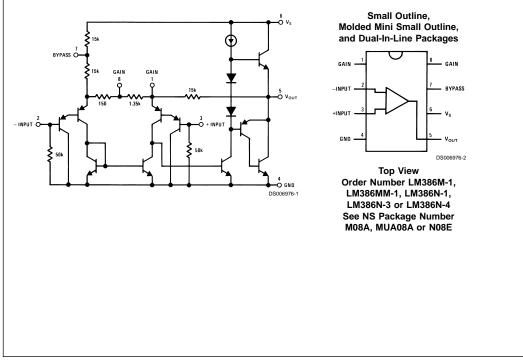
Features

- Battery operation
- Minimum external parts
- Wide supply voltage range: 4V-12V or 5V-18V
- Low quiescent current drain: 4 mA
- Voltage gains from 20 to 200
- Ground referenced input
- Self-centering output quiescent voltage
- Low distortion
- Available in 8 pin MSOP package

Applications

- AM-FM radio amplifiers
- Portable tape player amplifiers
- Intercoms
- TV sound systems
- Line drivers
- Ultrasonic drivers
- Small servo drivers
- Power converters

Equivalent Schematic and Connection Diagrams



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Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ecifications.	(SOIC and MSOP)	
	Vapor Phase (60 sec)	+215°C
15V	Infrared (15 sec)	+220°C
22V		
	on Product Reliability" for other meth surface mount devices.	nods of soldering
1.25W	Thermal Resistance	
0.73W	θ _{IC} (DIP)	37°C/W
0.595W		107°C/W
±0.4V	θ _{uc} (SO Package)	35°C/W
–65°C to +150°C	θ _{IA} (SO Package)	172°C/W
0°C to +70°C		210°C/W
+150°C	θ _{IC} (MSOP)	56°C/W
	22V 1.25W 0.73W 0.595W ±0.4V -65°C to +150°C 0°C to +70°C	$\begin{array}{c} \text{Vapor Phase (60 sec)} \\ \text{Vapor Phase (60 sec)} \\ 15\text{V} & \text{Infrared (15 sec)} \\ 22\text{V} & \text{See AN-450 "Surface Mounting Meth} \\ 0 & \text{Product Reliability" for other meth} \\ \text{surface mount devices.} \\ 1.25\text{W} & \text{Thermal Resistance} \\ 0.73\text{W} & \theta_{\text{JC}} (\text{DIP}) \\ 0.595\text{W} & \theta_{\text{JA}} (\text{DIP}) \\ \pm 0.4\text{V} & \theta_{\text{JC}} (\text{SO Package}) \\ -65^{\circ}\text{C to +150^{\circ}\text{C}} & \theta_{\text{JA}} (\text{SO Package}) \\ 0^{\circ}\text{C to +70^{\circ}\text{C}} & \theta_{\text{JA}} (\text{MSOP}) \end{array}$

Dual-In-Line Package

Soldering (10 sec)

Small Outline Package

+260°C

Electrical Characteristics (Notes 1, 2)

IA	=	25	C	

Parameter	Conditions	Min	Тур	Max	Units
Operating Supply Voltage (V _S)					
LM386N-1, -3, LM386M-1, LM386MM-1		4		12	V
LM386N-4		5		18	V
Quiescent Current (I _Q)	$V_{\rm S} = 6V, V_{\rm IN} = 0$		4	8	mA
Output Power (P _{OUT})					
LM386N-1, LM386M-1, LM386MM-1	$V_{S} = 6V, R_{L} = 8\Omega, THD = 10\%$	250	325		mW
LM386N-3	V_{S} = 9V, R_{L} = 8 Ω , THD = 10%	500	700		mW
LM386N-4	$V_{S} = 16V, R_{L} = 32\Omega, THD = 10\%$	700	1000		mW
Voltage Gain (A _V)	$V_{\rm S}$ = 6V, f = 1 kHz		26		dB
	10 µF from Pin 1 to 8		46		dB
Bandwidth (BW)	V _S = 6V, Pins 1 and 8 Open		300		kHz
Total Harmonic Distortion (THD)	$V_{\rm S} = 6V, R_{\rm L} = 8\Omega, P_{\rm OUT} = 125 \text{ mW}$		0.2		%
	f = 1 kHz, Pins 1 and 8 Open				
Power Supply Rejection Ratio (PSRR)	$V_{\rm S}$ = 6V, f = 1 kHz, $C_{\rm BYPASS}$ = 10 μ F		50		dB
	Pins 1 and 8 Open, Referred to Output				
Input Resistance (R _{IN})			50		kΩ
Input Bias Current (I _{BIAS})	V_{S} = 6V, Pins 2 and 3 Open		250		nA

Note 1: All voltages are measured with respect to the ground pin, unless otherwise specified.

Note 2: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which guarantee specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not guaranteed for parameters where no limit is given, however, the typical value is a good indication of device performance.

Note 3: For operation in ambient temperatures above 25°C, the device must be derated based on a 150°C maximum junction temperature and 1) a thermal resistance of 80°C/W junction to ambient for the dual-in-line package and 2) a thermal resistance of 170°C/W for the small outline package.

Application Hints

GAIN CONTROL

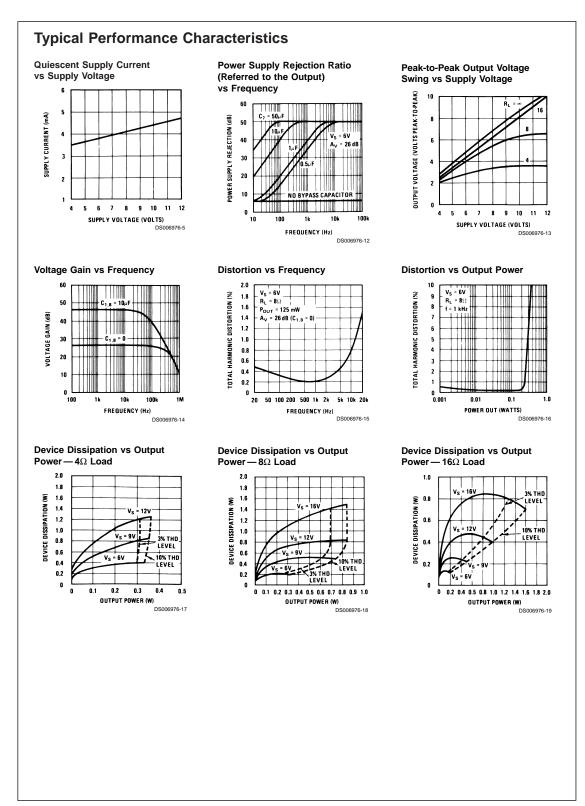
To make the LM386 a more versatile amplifier, two pins (1 and 8) are provided for gain control. With pins 1 and 8 open the 1.35 k Ω resistor sets the gain at 20 (26 dB). If a capacitor is put from pin 1 to 8, bypassing the 1.35 k Ω resistor, the gain will go up to 200 (46 dB). If a resistor is placed in series with the capacitor, the gain can be set to any value from 20 to 200. Gain control can also be done by capacitively coupling a resistor (or FET) from pin 1 to ground.

Additional external components can be placed in parallel with the internal feedback resistors to tailor the gain and frequency response for individual applications. For example, we can compensate poor speaker bass response by frequency shaping the feedback path. This is done with a series RC from pin 1 to 5 (paralleling the internal 15 k Ω resistor). For 6 dB effective bass boost: R \equiv 15 k Ω , the lowest value for good stable operation is R = 10 k Ω if pin 8 is open. If pins 1 and 8 are bypassed then R as low as 2 k Ω can be used. This restriction is because the amplifier is only compensated for closed-loop gains greater than 9.

INPUT BIASING

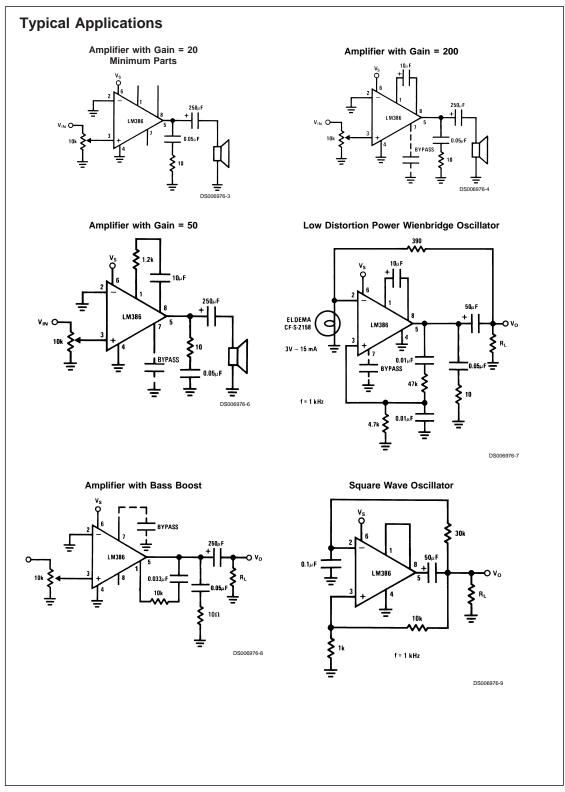
The schematic shows that both inputs are biased to ground with a 50 k Ω resistor. The base current of the input transistors is about 250 nA, so the inputs are at about 12.5 mV when left open. If the dc source resistance driving the LM386 is higher than 250 k Ω it will contribute very little additional offset (about 2.5 mV at the input, 50 mV at the output). If the dc source resistance is less than 10 k Ω , then shorting the unused input to ground will keep the offset low (about 2.5 mV at the input, 50 mV at the output). For dc source resistances between these values we can eliminate excess offset by putting a resistor from the unused input to ground, equal in value to the dc source resistance. Of course all offset problems are eliminate if the input is capacitively coupled.

When using the LM386 with higher gains (bypassing the 1.35 k Ω resistor between pins 1 and 8) it is necessary to bypass the unused input, preventing degradation of gain and possible instabilities. This is done with a 0.1 μF capacitor or a short to ground depending on the dc source resistance on the driven input.

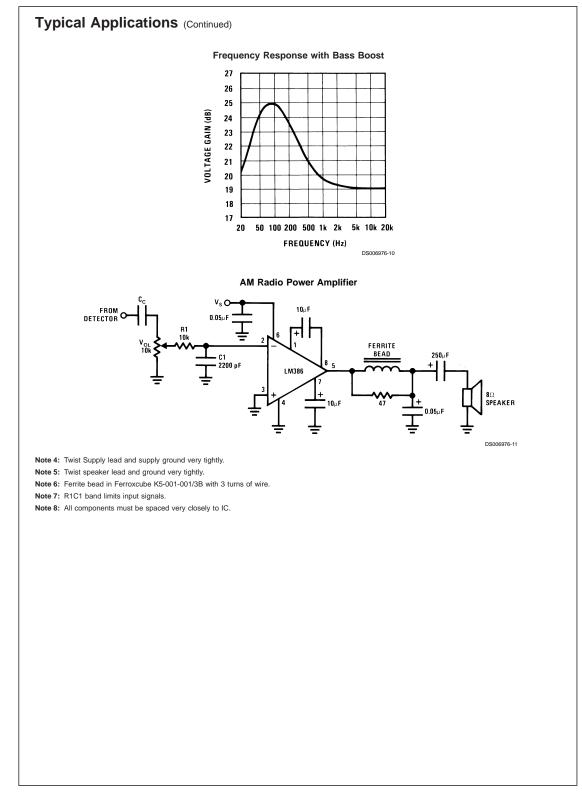


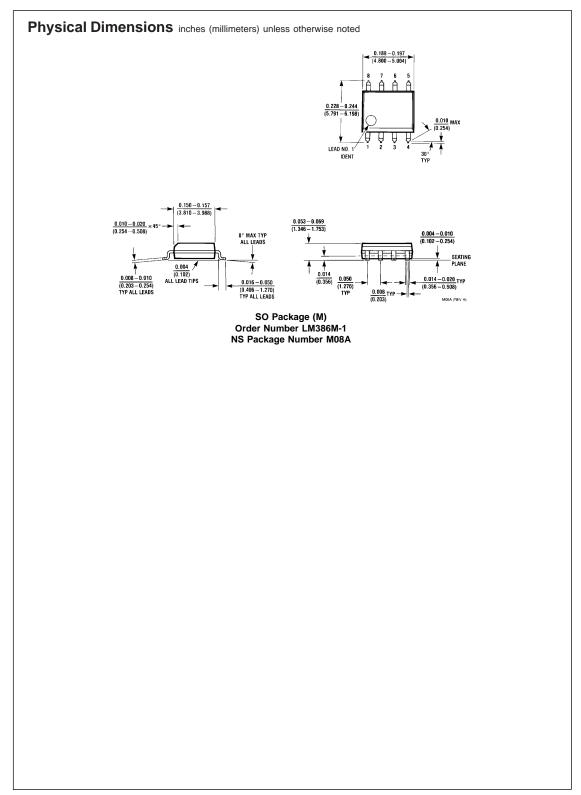
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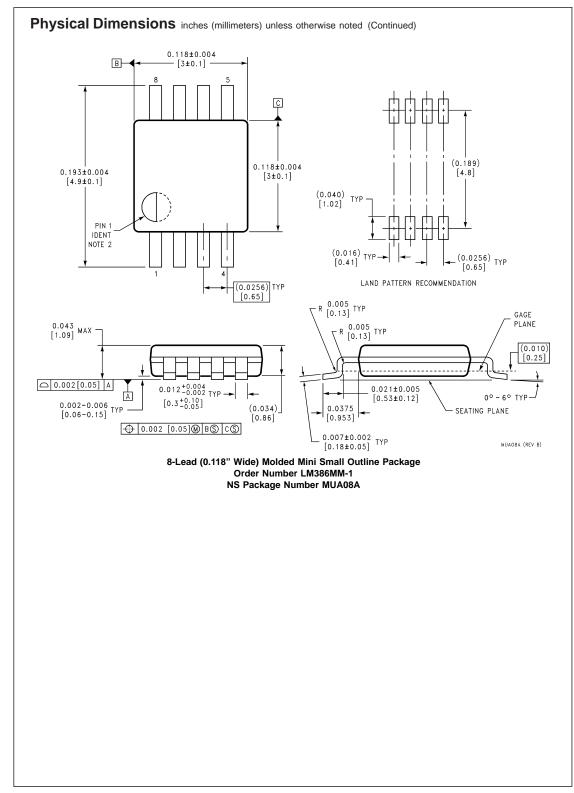


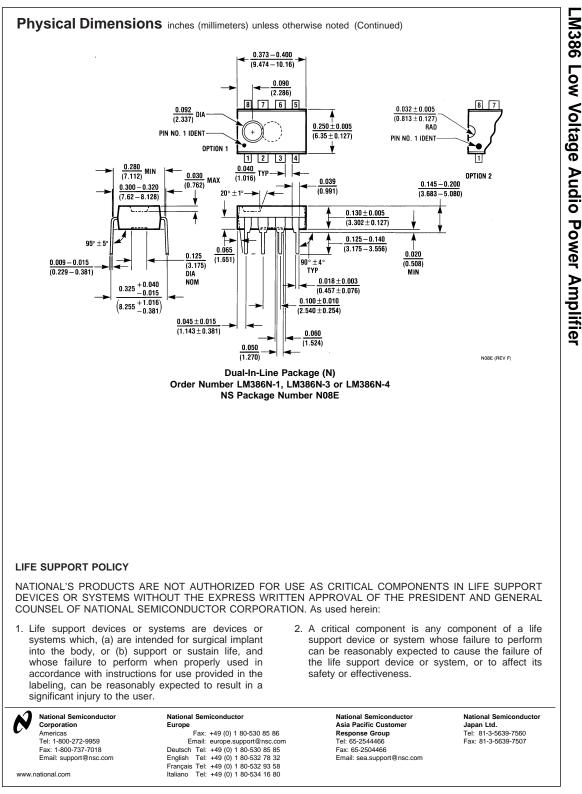
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